



State-Space Descriptions

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Chapter 2.9-2.10, 3.1-3.3



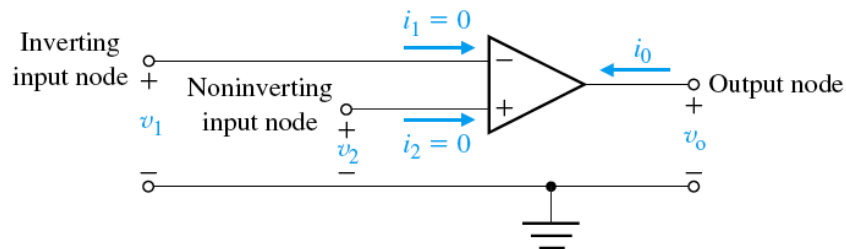
Outline

- Last class
 - Op-Amps; PID
 - Block diagram reductions
- Today
 - State differential equations
 - State-space representation
 - Transfer function of state-space form
- Next time
 - Linear Algebra needed to solve state-space eq'ns



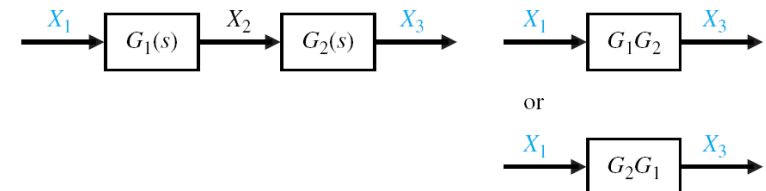
Review: Op-Amps

- Ideal operating conditions:
 - $i_1=0$ and $i_2=0$ (input impedance is infinite)
 - $V_1=V_2$



Block Diagram Transformations

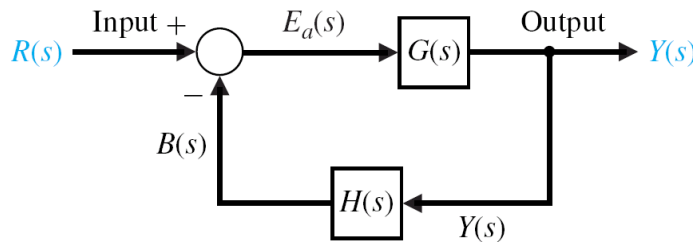
- Combining blocks in cascade: multiplication is commutative
- **When combining blocks, the input-output relationship (transfer function) should not change.**



$$X_3 = G_2 X_2 = G_2 (G_1 X_1) = G_2 G_1 X_1$$



Closed-Loop Transfer Function**



$$\frac{Y(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)}$$

Extremely important!

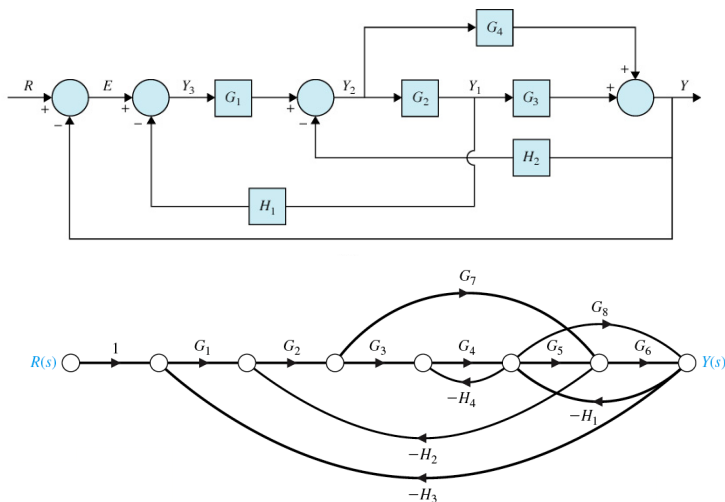


Signal Flow Graphs (SFGs)

- Alternative to block diagrams
- Block diagrams require iterative reduction to compute overall transfer function
- With SFG's, possible to compute overall transfer function using a formula, Mason's gain formula
- Dorf and Bishop, Chapter 2.7

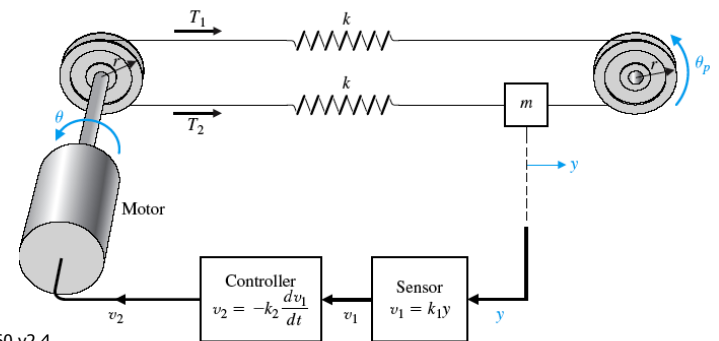


Signal Flow Graphs



Example: Printer belt-drive

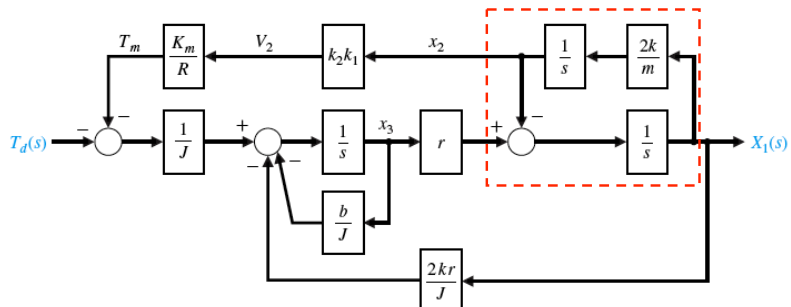
- Belt drive moves printing device
- Light sensor measures position of printing device
- Motor (actuator) adjusts belt drive





Example: Printer belt-drive

- Transfer function between
 - Output = position of printing device
 - Input = disturbance torques acting on the motor
- How can this be reduced?



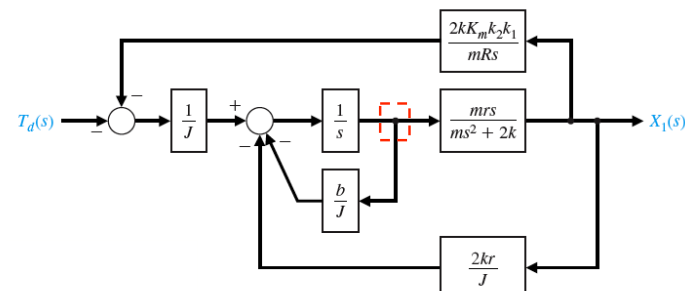
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Example: Printer belt-drive

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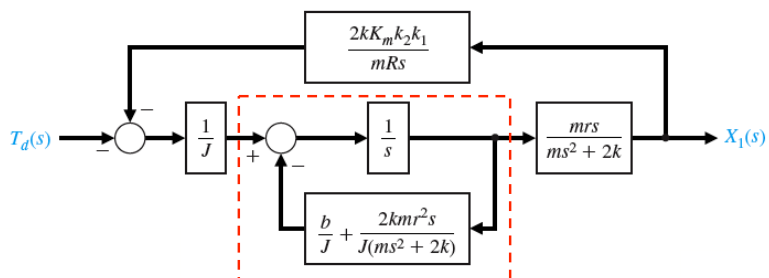
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Example: Printer belt-drive

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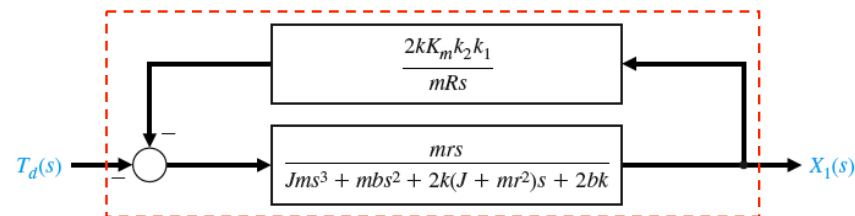
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Example: Printer belt-drive

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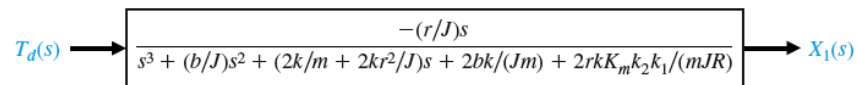
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Example: Printer belt-drive

- Transfer function between
 - Output = position of printing device
 - Input = disturbance torques acting on the motor
- How can this be reduced?
- Notes:
 - More than one way to achieve this reduction
 - Could also solve this algebraically from 1st diagram



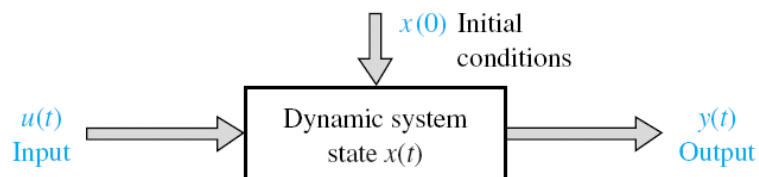
State equations

- State differential equations are an alternative way to describe a dynamic system (i.e. time-domain method)
- For LTI systems, it is routine to move between state-space and transfer function representations (i.e. between time-domain and frequency domain)
- Most multivariable design methods are based on state equations
- Basic technology: A physical system can be described by a high order differential equation



A General Dynamic System

- The state of a system is described in terms of state variables $[x_1(t), x_2(t), x_3(t), \dots, x_n(t)]$
- Given the input $u(t)$, initial conditions $x_0(t)$, and the present state $x(t)$, we can determine
 - the future system behavior
 - the output



State Variables of a Dynamic System

- “The state of a system is a set of variables such that the knowledge of these variables and of the input function will, **with the equations describing the dynamics**, provide the future state and output of the system”
- “The state variables describe the future response of a system, given the present state, the excitation inputs and the equations of the system”



The Concept of State

- Not new!
- Tycho Brahe, Kepler, Newton,...
- To predict the future motion of a planet, it is enough to know its current position and velocity as well as the equations governing its motion.
- Russian control school: Lyapunov, Pontryagin, etc. always worked in time domain
- The state is the minimum information required about a system to predict its future



The Concept of State

- The state gives a complete description of the system at a given time t , i.e. captures the evolution of the system up to time t
- This implies that there are orderly rules for transition from one state to another
- **State-space descriptions are not unique** – but input/output relations stay the same



Example: Spring-Mass-Damper

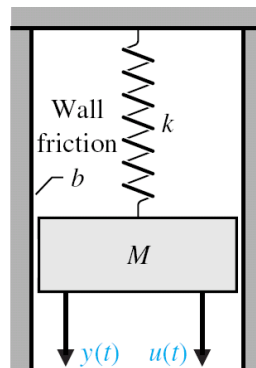
- Previously, we found $u(t) = M \frac{d^2 y(t)}{dt^2} + b \frac{dy(t)}{dt} + ky(t)$
- Now define $x_1(t) = y(t), x_2(t) = \frac{dy(t)}{dt}$
- such that

$$u(t) = M \frac{dx_2(t)}{dt} + bx_2(t) + kx_1(t)$$

- ** ■ Write as 2 first-order diff. eqns.

$$\frac{dx_1(t)}{dt} = x_2(t)$$

$$\frac{dx_2(t)}{dt} = -\frac{b}{M}x_2(t) - \frac{k}{M}x_1(t) + \frac{1}{M}u(t)$$



The Concept of State

- Typically for a physical system, the state relates to the concept of energy – each state describes an energy storage component (mass, spring, inductor, capacitor etc.)
- Often it is useful to use variables that are readily measured, e.g. currents, voltages, positions, velocities, pressures, temperatures, concentrations, etc...

- State vector $x(t)$:

$$\mathbf{x}(t) = \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_n(t) \end{bmatrix}; \quad \dot{\mathbf{x}}(t) = \begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \\ \vdots \\ \dot{x}_n(t) \end{bmatrix};$$



State-Space Description

- The state-space description leads to a system of coupled, first-order differential equations
- Reduction to first-order is a general technique to transform a high-order differential equation into a state-space model

$$y(t) = a_0 f(t) + a_1 f'(t) + a_2 f''(t) + \dots + a_n f^{(n)}(t)$$

$$\begin{aligned} x_1'(t) &= a_{11}x_1(t) + a_{12}x_2(t) + \dots + a_{1n}x_n(t) \\ x_2'(t) &= a_{21}x_1(t) + a_{22}x_2(t) + \dots + a_{2n}x_n(t) \\ &\vdots \\ x_n'(t) &= a_{n1}x_1(t) + a_{n2}x_2(t) + \dots + a_{nn}x_n(t) \end{aligned}$$



State-Space Description

- Write in matrix form

$$\begin{aligned} x_1'(t) &= a_{11}x_1(t) + a_{12}x_2(t) + \dots + a_{1n}x_n(t) \\ x_2'(t) &= a_{21}x_1(t) + a_{22}x_2(t) + \dots + a_{2n}x_n(t) \\ &\vdots \\ x_n'(t) &= a_{n1}x_1(t) + a_{n2}x_2(t) + \dots + a_{nn}x_n(t) \end{aligned}$$

$$\dot{x} = Ax, \quad x = \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_n(t) \end{bmatrix}, \quad A = \begin{bmatrix} a_{11} & a_{12} & \dots & a_{1n} \\ a_{21} & a_{22} & \dots & a_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ a_{n1} & a_{n2} & \dots & a_{nn} \end{bmatrix}$$



The State Differential Equation

- The general state space description for a *linear time-invariant, continuous-time* dynamical system is:

State differential equation: $\dot{x}(t) = Ax(t) + Bu(t)$ (1)

Output equation: $y(t) = Cx(t) + Du(t)$ (2)

Labels: *system matrix* (A), *input matrix* (B), *input vector* (u), *output matrix* (C)

where $x(t) \in \mathbb{R}^n$, $u(t) \in \mathbb{R}^m$ and $y(t) \in \mathbb{R}^p$.

- A is $(n \times n)$, B is $(n \times m)$, C is $(p \times n)$ and D is $(p \times m)$. Shorthand for this system is $[A, B, C, D]$



Example: RLC Circuit

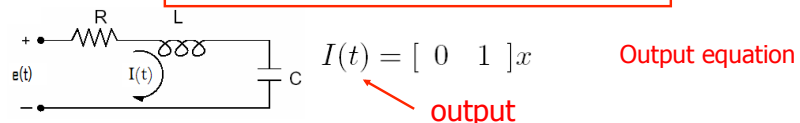
- Consider an RLC circuit described by:

$$e(t) = RI(t) + L \frac{d}{dt} I(t) + \frac{1}{C} \int_0^t I(\tau) d\tau$$

With $x_1(t) = \int_0^t I(\tau) d\tau$ and $x_2(t) = I(t)$ the state-space formulation is

$$\frac{d}{dt} x = \begin{bmatrix} 0 & 1 \\ -\frac{1}{LC} & -\frac{R}{L} \end{bmatrix} x + \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} e(t)$$

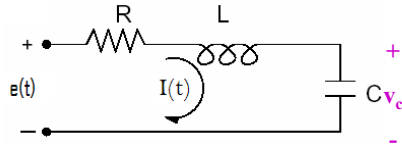
Labels: *state vector* (x), *State differential equation*





Example: RLC Circuit

- State-space representations are NOT unique



$$\begin{aligned} i &= x_1 & i &= C\dot{v}_c \\ v_c &= x_2 & \dot{x}_2 &= \frac{1}{C}x_1 \end{aligned}$$

$$e = iR + L \frac{di}{dt} + v_c$$

$$e = Rx_1 + L\dot{x}_1 + x_2$$

$$\dot{x}_1 = -\frac{R}{L}x_1 - \frac{1}{L}x_2 + \frac{1}{L}e$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -\frac{R}{L} & -\frac{1}{L} \\ \frac{1}{C} & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} e$$

$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

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Example: Spring-Mass-Damper

- First-order differential equations

$$\frac{dx_1(t)}{dt} = x_2(t)$$

$$\frac{dx_2(t)}{dt} = -\frac{b}{M}x_2(t) - \frac{k}{M}x_1(t) + \frac{1}{M}u(t)$$

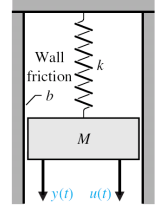
- In matrix form:

$$\begin{bmatrix} \frac{dx_1(t)}{dt} \\ \frac{dx_2(t)}{dt} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\frac{k}{M} & -\frac{b}{M} \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix} u(t)$$

$$y(t) = \begin{bmatrix} 1 & 0 \end{bmatrix} x(t)$$

- with

$$A = \begin{bmatrix} 0 & 1 \\ -\frac{k}{M} & -\frac{b}{M} \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix}, \quad C = \begin{bmatrix} 1 & 0 \end{bmatrix}, \quad D = 0$$



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Solution to the State Eq'ns

- Why is this representation so useful?
Because we know what its solution looks like.

- Given the state equations

$$\dot{x}(t) = Ax(t) + Bu(t)$$

$$y(t) = Cx(t) + Du(t)$$

- We can show that

$$x(t) = e^{At}x(0) + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau$$

****Note that this is a matrix exponential!**

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Transfer function of state eqns

- Take the Laplace transform of

$$\dot{x}(t) = Ax(t) + Bu(t)$$

$$y(t) = Cx(t) + Du(t)$$

- Re-arrange to find $X(s)$ in terms of $U(s)$, $x(0)$:

$$sX(s) - x(0) = AX(s) + BU(s)$$

$$(sI - A)X(s) = BU(s) + x(0)$$

$$\text{So, } X(s) = (sI - A)^{-1}BU(s) + (sI - A)^{-1}x(0)$$

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Transfer function of state eqns

- Assume $x(0)=0$ for simplicity.
- Then substituting

$$X(s) = (sI - A)^{-1}BU(s)$$

into the output equation

$$Y(s) = CX(s) + DU(s)$$

yields

$$Y(s) = (C(sI - A)^{-1}B + D)U(s) = G(s)U(s)$$

- For SISO case, the **transfer function** is

$$G(s) = Y(s)/U(s)$$



Summary

- Reduction from n^{th} order differential equation to n first order differential equations
- State-space representations are NOT UNIQUE
- Known closed-form solution
- Transfer function for SISO state-space description

- Next time:
 - Solving the state-space equations
 - Linear algebra review