## Response of LTI Systems Using Laplace Transforms

$$
H(s)=\int_{-\infty}^{\infty} h(t) e^{-s t} d t
$$

Where $h(t)$ is an impulse response, is called the system function or transfer function and it completely characterizes the input/output relationship of an LTI system. We can use it to determine time responses of LTI systems.

## Transfer Functions

We can use Laplace Transforms to solve differential equations for systems (assuming the system is initially at rest for one-sided systems) of the form:

$$
\sum_{k=0}^{n} a_{k} \frac{d^{k}}{d t^{k}} y(t)=\sum_{k=0}^{m} b_{k} \frac{d^{k}}{d t^{k}} x(t)
$$

Taking the Laplace Transform of both sides of this equation and using the Differentiation Property, we get:

$$
Y(s) \sum_{k=0}^{n} a_{k} s^{k}=X(s) \sum_{k=0}^{m} b_{k} s^{k}
$$

From this, we can define the transfer function, $\boldsymbol{H}(\boldsymbol{s})$, as:

$$
H(s)=\frac{Y(s)}{X(s)}=\frac{\sum_{k=0}^{m} b_{k} s^{k}}{\sum_{k=0}^{n} a_{k} s^{k}}
$$

Which is the ratio of two polynomials in " s "

## Partial Fraction Expansion

Instead of taking contour integrals to invert Laplace Transforms, we will use Partial Fraction Expansion. We review it here. Given a Laplace Transform,

$$
F(s)=\frac{b_{m} s^{m}+\cdots+b_{1} s+b_{0}}{s^{n}+a_{n-1} s^{n-1}+\cdots+a_{1} s+a_{0}}=\frac{N(s)}{D(s)}, \quad m<n
$$

If $m$ isn't less than $n$, perform polynomial division and then the remainder can be analyzed by Partial Fraction Expansion.
We write its Partial Fraction Expansion as:

$$
\begin{aligned}
F(s) & =\frac{N(s)}{\left(s-p_{1}\right)\left(s-p_{2}\right) \cdots\left(s-p_{n}\right)} \\
& =\frac{k_{1}}{s-p_{1}}+\frac{k_{2}}{s-p_{2}}+\cdots+\frac{k_{n}}{s-p_{n}}
\end{aligned}
$$

where

Assuming that all of the poles have unique values!

$$
k_{j}=\left.\left(s-p_{j}\right) F(s)\right|_{s=p_{j}}
$$

is the residue of the pole at $p j$.

Thus

$$
f(t)=\sum_{j=1}^{n} k_{j} e^{p_{j} t} u(t)
$$

because the Inverse Laplace Transform of

$$
\frac{k_{j}}{\left(s+p_{j}\right)} \text { is } k_{j} e^{p j^{z}}
$$

## Convolution

An important property of Laplace Transforms is that the Laplace transform of the convolution of two signals is the product of their Laplace transforms:

$$
x(t)^{*} h(t) \leftrightarrow X(s) H(s)
$$

This is useful for studying LTI systems. In fact, we can completely characterize an LTI system from:

1. The system differential equation or
2. the system transfer function $H(s)$ or
3. the system impulse response $h(t)$.

Example 1 Find $y(t)$ where the transfer function $H(s)$ and the input $x(t)$ are given. Use Partial Fraction Expansion to find the output $y(t)$ :

$$
H(s)=\frac{3 s+1}{s^{2}+6 s+5}, \quad x(t)=e^{-3 t} u(t)
$$

First find the Laplace transform of $x(t)$ from a table of Laplace transforms: $X(s)=1 /(s+3)$

Now find the Laplace Transform of the output by multiplying $H(s)$ by $X(s)$
$Y(s)=(3 s+1) /\left[(s+3)^{*}(s+2)^{*}(s+3)\right]$
To do partial fractions we need to separate out the poles, but one of the poles is repeated so we need to find
$Y(s)=A /(s+2)+B /(s+3)+C /(s+3)^{2}$

The third term is to account for the repeated root.

Multiplying both sides of the equation by ( $s+2$ ) yields:
$(s+2)^{*} Y(s)=(3 s+1) /[(s+3) *(s+3)]=A+(s+2) *\left[B /(s+3)+C /(s+3)^{2}\right]$

Letting s ---> -2 now yields
$-5 /[(1) *(1)]=A$ so $\mathbf{A}=-\mathbf{5}$

If we now repeat the process but multiply both sides by $(s+3)^{2}$ we get:
$(s+3)^{2 *} Y(s)=(3 s+1) /[(s+2)]=A^{*}(s+3)^{2}+B^{*}(s+3)+C$
Now take the limit as s ---> -3 and
$(-9+1) /(-1)=C$ so $\mathbf{C}=\mathbf{8}$
there are two methods to find the remaining constant, B

1. Take ${ }^{2}$ the derivative of both sides of the original equation and then $B$ can be isolated by multiplying both sides by $(s+3)$ and taking the limit as $s-->-3$. This is the method shown in most textbooks.
2. go back to the original equation setting $A=1$ and $C=8$. Then put the three terms over a common denominator and the extra terms should cancel out to leave the " $B$ " term.
Using the second method here:
$Y(s)=-5 /(s+2)+8 /(s+3)+B /(s+3)$
$Y(s)=\left[-5(s+3)^{2}+8(s+2)+B(s+2)(s+3)\right] /\left[(s+2)(s+3)^{2}\right]$
$Y(s)=\left[-5 s^{2}-30 s-45-+8 s+16+B\left(s^{2}+5 s+6\right)\right] /\left[(s+2)(s+3)^{2}\right]$
$Y(s)=(B-5) s^{2}+(5 B-22) s+(6 B-45) /\left[(s+2)(s+3)^{2}\right]$
But $Y(s)=(3 s+1) /\left[(s+3)^{*}(s+2)^{*}(s+3)\right]$ so
$B-5=0$ or $\mathbf{B}=\mathbf{5}$ from the $s^{2}$ term but as a check from the $s^{1}$ term:
$5 B-22=3$ OK and from the $s^{0}$ term
$6 \mathrm{~B}-29=1$ OK

We Have the partial fraction expansion of:
$Y(s)=-5 /(s+2)+5 /(s+3)+8 /(s+3)^{2}$
And using the Table of Laplace Transforms
$y(t)=\left[-5 * \exp (-2 t)+5^{*} \exp (-3 t)+8 t^{*} \exp (-3 t)\right] * U(t)$

## Stability

We saw that a condition for bounded-input bounded-output stability was:

$$
\int_{-\infty}^{\infty}|h(t)| d t<\infty
$$

Let's look at stability from a system function standpoint.
Given a Laplace Transform $H(s)$, we expand $H(s)$ with Partial Fraction Expansion:

$$
H(s)=\frac{k_{1}}{s-p_{1}}+\frac{k_{2}}{s-p_{2}}+\cdots+\frac{k_{n}}{s-p_{n}}
$$

The corresponding impulse response is:

$$
h(t)=k_{1} e^{p_{1} t} u(t)+k_{2} e^{p_{2} t} u(t)+\cdots+k_{n} e^{p_{n} t} u(t)
$$

What happens to $h(t)$ as $t \rightarrow \infty$ ? For a system to be stable, its impulse response must not blow up as $t \rightarrow \infty$.

If $\operatorname{Re}\left\{p_{i}\right\}, \forall i$, then $h(t)$ decays to 0 as $t \rightarrow \infty$ and the system is stable (just evaluate $\int_{-\infty}^{\infty}|h(t)| d t$ ).
Therefore, the system is BIBO stable if and only if all poles of $H(s)$ are in the left half plane of the $s$-plane.

